

AMENDMENTS TO THE CLAIMS:

This listing of claims will replace all prior versions and listings of claims in the application:

1. (Currently Amended) A method for controlling a synchronous permanent magnet multiple-phase motor, the motor having multiple phases and having a rotor, the method comprising ~~the steps of~~:

[[-]] controlling [[the]] drive ~~current~~ currents supplied to [[each]] the phases of the motor [[phase]] by turning [[it]] the drive currents off at a predetermined frequency[[,]];

[[-]] measuring, at said predetermined frequency, [[the]] induced voltages of at least two of the [[motor]] phases of the motor, when the [[power]] drive currents in said at least two of the phases of the motor phase is are turned off, with a sensitivity ~~allowing to obtain~~ sufficient for obtaining significant voltages voltage values at a near-zero speed of the rotor[[,]]

[[-]] determining ~~the rotor~~ a position and/or a the rotor speed of the rotor from ~~signals resulting from said measured induced voltages~~[[,]];

[[-]] ~~entering said determined rotor~~ filtering the position and/or the said- ~~determined rotor speed of the rotor~~ [[into]] with a state filter ~~which delivers to obtain a~~ filtered rotor position and/or a filtered rotor speed[[,]]; and

[[-]] adjusting the drive ~~current as a function of said~~ currents according to the filtered rotor position and/or the filtered rotor speed.

2. (Currently Amended) The method according to claim 1, wherein the state filter is ~~arranged~~ configured so as to take into account that when the speed of the ~~[[motor]]~~ rotor is very low, the position of the rotor ~~[[can]]~~ does not change substantially over a short period of time.

3. (Original) The method according to claim 2, wherein said state filter is a Kalman filter.

4. (Currently Amended) The method according to any of claims 1 to 3, wherein the ~~measured~~ position, θ , of the rotor is determined by the formula

$$\theta = \arctg\left(\frac{U_a}{U_b}\right),$$

where U_a is ~~equal to~~ the measured induced voltage in one of the phases of the motor ~~[[phase]]~~ and U_b is equal to $\frac{V_2 - V_3}{\sqrt{3}}$, V_2 and V_3 being the measured induced voltages in two others of the other two phases of the motor.

5. (Currently Amended) The method according to ~~any of claims~~ claim 1 ~~[[to 4]]~~, wherein the ~~measured~~ speed of the rotor is determined by computing ~~[[the]]~~ a square root of ~~[[the]]~~ a sum of squares of the measured induced voltages.

6. (Currently Amended) The method according to ~~any of claims~~ claim 2 ~~[[to 5]]~~, wherein the ~~[[said]]~~ state filter implements an algorithm such as

$$X = X_{-1} + (a * V * T + b * dP) \div 2,$$

where X is ~~[[the]]~~ an estimated position of the rotor at time t ,
 X_{-1} is ~~[[the]]~~ an estimated position of the rotor at time t_{-1} ,
 ~~X_m is the measured position using back EMF voltages at time t_m with $(t_{-1} \leq t_m \leq$~~
 ~~$t)$,~~
 V is ~~[[the]]~~ a measured speed of the rotor using back EMF voltages at time t_m ,
 T is ~~[[the]]~~ a time duration between 2 successive measurements ~~$(t_{-1}$ and $t[[]])$,~~
 dP is the difference between X_m and X_{-1} , wherein X_m is a measured position
of the rotor using back EMF voltages at time t_m with $t_{-1} \leq t_m \leq t$, and the ~~[[such]]~~
difference dP is being however limited to $\pm(c * VT + d)$, and
 a , b , c and d are coefficients which depend on characteristics of the motor-
characteristics.

7. (Currently Amended) An electronic device for controlling a synchronous permanent magnet motor ~~[[(1)]]~~ with at least one phase, a coil, a rotor, and a motor driver ~~[[(2)]]~~, the electronic device comprising:

detection means ~~(3), which are~~ connected to the at least one phase ~~phases (A, B, C)~~ of the motor and deliver for generating signals that represent induced voltages of the at least one phase of the motor ~~phases~~, said detection means having a ~~high enough~~ gain ~~to provide significant output~~ such that the signals representing the induced voltages are significant even if ~~[[the]]~~ a speed of the rotor is near-zero~~[[,]]~~; and

a control circuit ~~[[(4)]]~~ connected to said detection means and to the motor driver ~~(2), which supplies~~ for supplying driving currents to the motor, said control circuit comprising means for generating signals representing a ~~computing the~~ position and/or

~~[[the]]~~ a speed of the rotor from the ~~output signals provided by said detection means~~
representing the induced voltages.

8. (Currently Amended) The electronic device according to claim 7, wherein
the motor includes at least two phases, and wherein said detection means ~~comprise~~
comprises, for each of the at least two phases ~~phase of the motor;~~

a differential amplifier ~~(31, 32, 33)~~ the having inputs ~~of which are~~ connected to
two of the at least two ~~different~~ phases of the motor; and

an analog-to-digital converter ~~(34, 35, 36)~~ to convert the for converting an analog
signal outputted by said differential amplifier into a digital signal, ~~which is applied and~~
providing said digital signal to said control circuit.

9. (Currently Amended) The electronic device according to claim 7 or claim
8, wherein the control circuit further comprises a state filter for filtering the signals
representing the position and/or the speed of the rotor ~~motor determined from the output~~
~~signals of said detection means.~~

10. (Currently Amended) The electronic device according to claim 9, wherein
said state filter is a Kalman filter.

11. (New) The method according to claim 4, wherein the speed of the rotor is
determined by computing a square root of a sum of squares of the measured induced
voltages.

12. (New) The method according to claim 4, wherein the state filter implements an algorithm such as

$$X = X_{-1} + (a * V * T + b * dP) \div 2,$$

where X is an estimated position of the rotor at time t ,

X_{-1} is an estimated position of the rotor at time t_1 ,

V is a measured speed of the rotor using back EMF voltages at time t_m ,

T is a time duration between t_1 and t ,

dP is the difference between X_m and X_{-1} , wherein X_m is a measured position of the rotor using back EMF voltages at time t_m with $t_1 \leq t_m \leq t$, and the difference dP is limited to $\pm(c * VT + d)$, and

a , b , c and d are coefficients which depend on characteristics of the motor.

13. (New) The method according to claim 5, wherein the state filter implements an algorithm such as

$$X = X_{-1} + (a * V * T + b * dP) \div 2,$$

where X is an estimated position of the rotor at time t ,

X_{-1} is an estimated position of the rotor at time t_1 ,

V is a measured speed of the rotor using back EMF voltages at time t_m ,

T is a time duration between t_1 and t ,

dP is the difference between X_m and X_{-1} , wherein X_m is a measured position of the rotor using back EMF voltages at time t_m with $t_1 \leq t_m \leq t$, and the difference dP is limited to $\pm(c * VT + d)$, and

a , b , c and d are coefficients which depend on characteristics of the motor.